

# Homework 6

## 1. Problem:

(a) Using the figures on pages 133 and 134 of the Lecture Notes, Part II, as a guide, sketch the band diagram of an MOS capacitor with an  $n$ -type Si substrate in **i)** accumulation, **ii)** at flat band condition, and **iii)** at the onset of strong inversion (that is,  $\psi_s = 2\psi_B$ ). Assume an Al gate with electron affinity  $e\phi_M = 3.0$  eV, use for Si the work-function  $e\chi = 3.2$  eV, and assume the  $n$ -type substrate to be doped with  $N_D = 3 \times 10^{17}$  donors/cm<sup>3</sup>.

(b) What is the value of the flatband voltage  $V_{FB}$  and of the threshold voltage  $V_{T0}$  (see Eq. (247) of the Notes, Part III)?

(c) What is the value of the depletion capacitance  $C_D$  at the onset of strong inversion?

## Solution:

(a) See the figures.

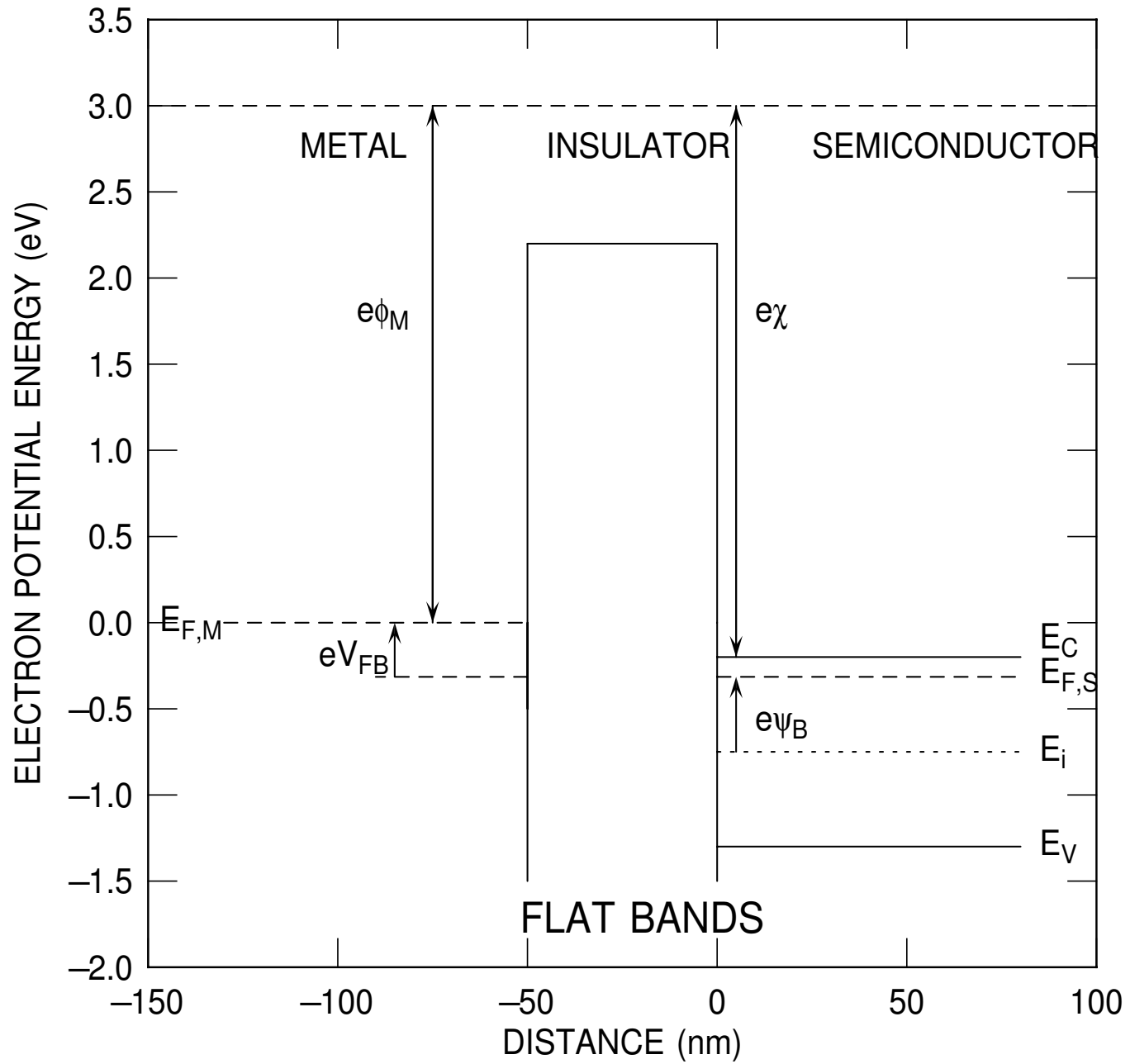
(b): Looking at the first figure, the distance between the metal Fermi level and the vacuum level (the topmost dashed horizontal line) is given by:

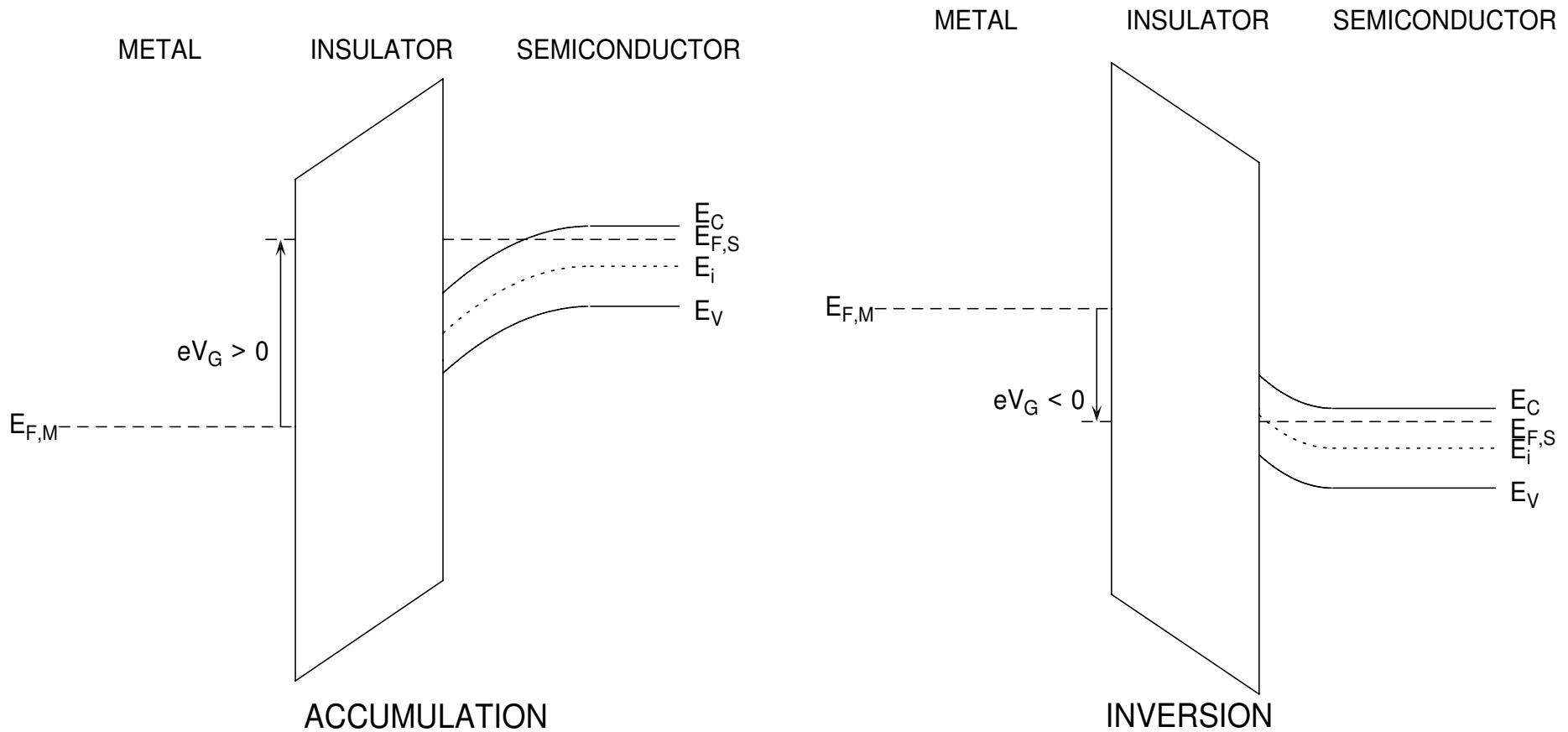
$$|V_{FB}| + e\phi_M = |V_{FB}| + 3.0 \text{ eV} . \quad (1)$$

Similarly, the distance between the Si Fermi level and the vacuum level will be given by:

$$e\chi + E_i - \psi_B = 3.2 + 0.55 - 0.435 \text{ eV} = 3.315 \text{ eV} , \quad (2)$$







where  $\psi_B = k_B T \ln(N_D/n_i)$  is the distance between the Fermi level and the intrinsic level  $E_i$ . From Eqns. (1) and (2), noticing that  $V_{FB}$  must be negative (since it raises the potential energy for electrons in the gate), we have  $V_{FB} = -0.315 \text{ V}$ .

Finally, the threshold voltage is given by Eq. (247) of the Notes with  $\gamma = (2\epsilon_{Si}eN_D)^{1/2}$  (with  $N_D$  replacing  $N_A$  since we are dealing with an  $n$ -type substrate. Thus, using  $C_{ox} = \epsilon_{ox}/t_{ox}$  and  $t_{ox} = 5 \text{ nm}$  (from Problem 2):

$$C_{ox} = \frac{11.7 \times 8.85 \times 10^{-12} \text{ F/m}}{5 \times 10^{-9} \text{ m}} = 6.9 \times 10^{-3} \text{ F/m}^2. \quad (3)$$



To compute the body-factor  $\gamma$  let's convert the doping density to  $m^{-3}$ :

$$\gamma = \frac{(2 \times 1.6 \times 10^{-19} \times 11.7 \times 8.85 \times 10^{-12} \times 3 \times 10^{23})^{1/2}}{6.9 \times 10^{-4}} \approx \frac{9.22 \times 10^{-4}}{6.9 \times 10^{-3}} = 0.134 \text{ V}^{1/2} \quad (4)$$

Thus:

$$V_{T0} = V_{FB} + 2\psi_B + \gamma (2\psi_B)^{1/2} \approx -0.315 + 0.87 + 0.125 = 0.68 \text{ V} \quad (5)$$

(c) The depletion capacitance is given by:

$$C_D = \frac{\epsilon_{Si}}{W_D}, \quad (6)$$

where  $W_D = [2\epsilon_{Si}\psi_s/(eN_D)]^{1/2}$  is the depletion width. The minimum capacitance is obtained when the depletion width reaches its maximum value, for  $\psi_s = 2\psi_B$ . Using the value for  $\psi_B$  calculated above we have:

$$W_{D,max} = \left( \frac{4\epsilon_{Si}\psi_B}{eN_D} \right)^{1/2} \approx 61.3 \text{ nm}. \quad (7)$$

So:

$$C_{D,min} = \frac{11.7 \times 8.85 \times 10^{-12} \text{ F/m}}{6.13 \times 10^{-8} \text{ m}} = 1.7 \times 10^{-7} \text{ F/cm}^2 \quad (8)$$

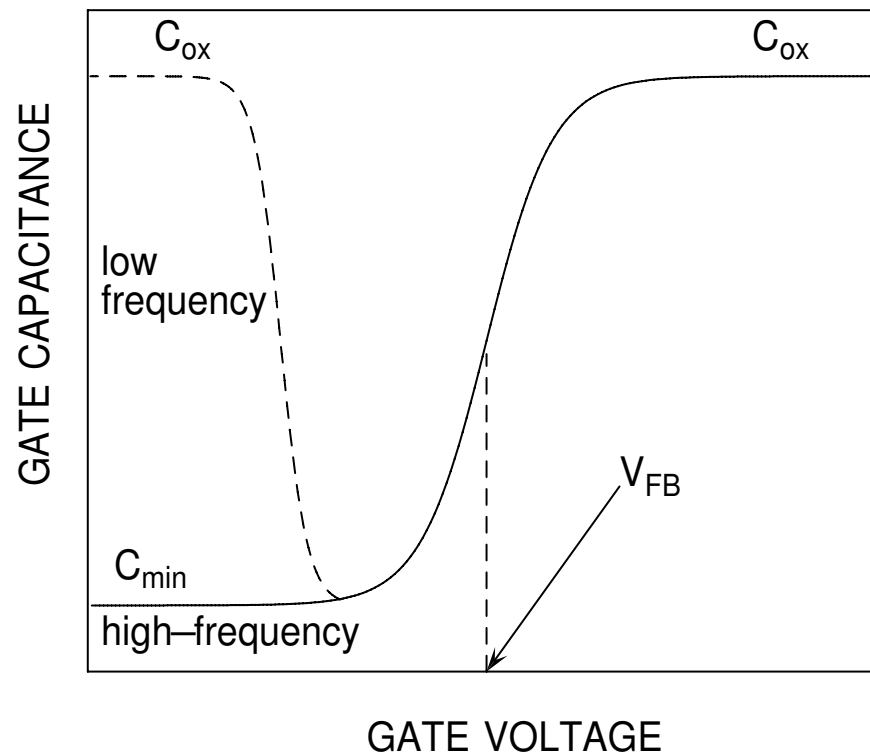
## 2. Problem:

Assuming an oxide thickness  $t_{ox}$  of 5 nm and a gate area of  $100 \mu\text{m}^2$ , sketch the capacitance-voltage ( $C - V$ ) characteristics of the MOS capacitor of problem 1, as in the figure at page 140 of the Lecture Notes. Recall that this capacitor is on an  $n$ -type substrate!

## Solution:

See the following figure:





Note that we have already calculated  $C_{ox} \approx 6.9 \times 10^{-3} \text{ F/m}^2$  and  $C_D \approx 1.7 \times 10^{-3} \text{ F/m}^2$ . Since the area of the capacitor is  $100 \mu\text{m}^2 = 10^{-6} \text{ cm}^2$ , we have  $C_{ox} \approx 0.69 \text{ pF}$ ,  $C_D \approx 0.17 \text{ pF}$ , so that the minimum capacitance at high frequency in inversion will be  $C_{min} = 1/(1/C_{ox} + 1/C_D) \approx 0.14 \text{ pF}$ .

### 3. Problem:

Consider a Si  $n$ MOSFET with the following parameters:

substrate doping ( $p$ -type):  $N_A = 5 \times 10^{16} \text{ cm}^{-3}$

channel length  $L = 0.5 \mu\text{m}$

gate width  $W = 5 \mu\text{m}$

electron mobility  $\mu_n = 600 \text{ cm}^2/\text{Vs}$

oxide thickness  $t_{ox} = 15 \text{ nm}$

flatband voltage  $V_{FB} = 0$ .

Using the simplified model of the Lecture Notes, pages 159-162, especially Eq. (255), plot as accurately as you



can the  $I_D - V_D$  characteristics of the device assuming that the source contact is grounded (that is,  $V_S=0$ ). More specifically, plot  $I_D$ -vs- $V_D$  for  $V_G - V_{T0} = 0.0, 0.5, 1.0, 1.5,$  and  $2.0$  V for  $V_D$  ranging from 0 to 5 V. Indicate as best as you can the separation between the linear and saturated region by computing  $V_{D,sat}$  for the various values of  $V_G - V_{T0}$ . Also indicate the value of the threshold voltage  $V_{T0}$ .

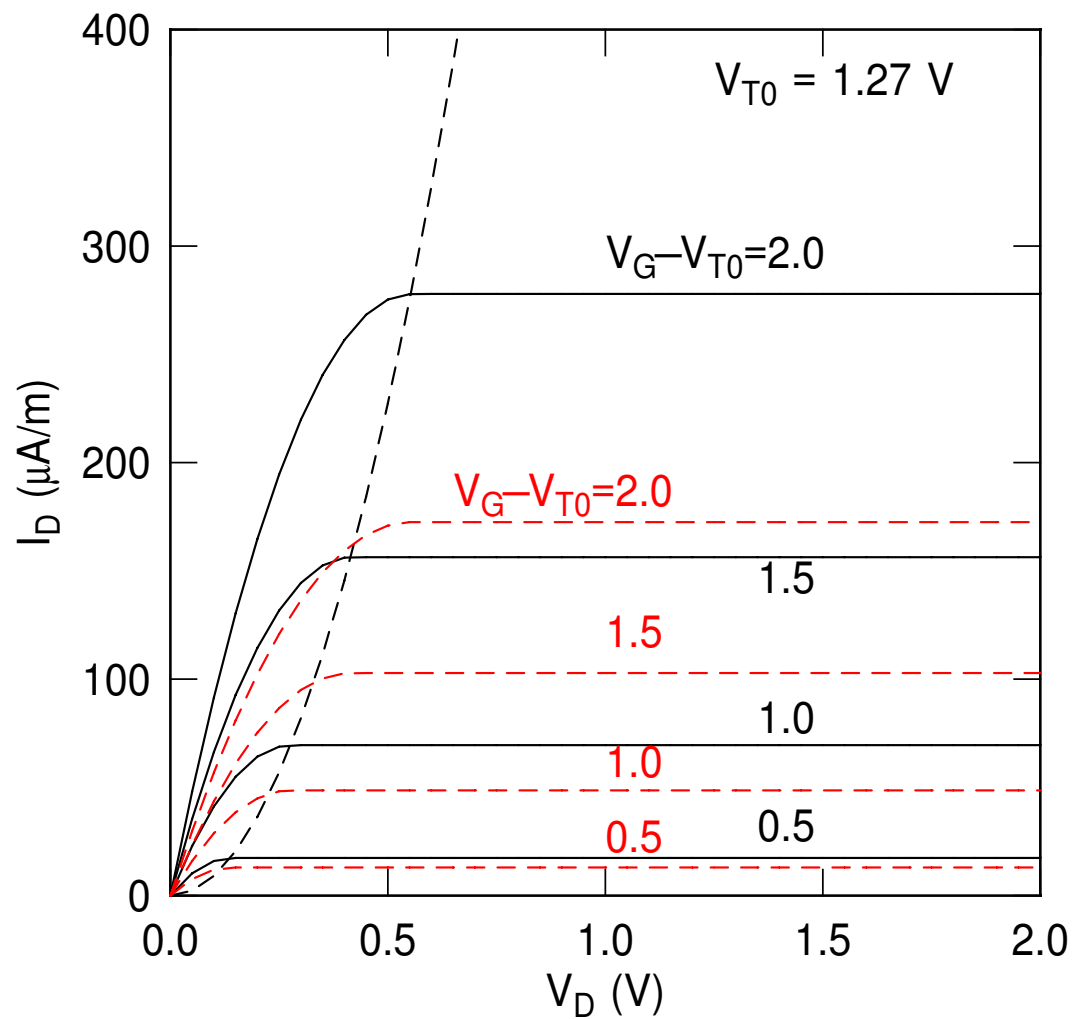
**Solution:**

The figure below shows the drain current as a function of drain bias in four curves (black, solid lines) parametrized by the gate overdrive  $V_G - V_{T0}$ . Note that  $V_{T0} = 1.27$  V and  $n = 1.32$ . The parabola indicated by a dashed black line is given by the expression:

$$I_D = \mu C_{ox} \frac{W}{L} \frac{n}{2} V_D^2, \quad (9)$$

and it separates the linear (left) from the saturated region (right). When  $V_D$  exceeds  $V_{D,sat}$  – so, at the right of the dashed parabola – Eq. (255) is not valid, as explained at page 160, immediately after Eq. (255), because the channel is pinched off. Many of you, especially those who have employed a computer to evaluate numerically the current, have forgotten this fact and have used Eq. (255) also when  $V_D > V_{D,sat}$ , getting negative values for the current. Not good...





4. **Problem:**

As in problem 3 above, plot the same  $I_D - V_D$  characteristics, but now account for the degradation of the electron mobility with increasing  $V_G$  via Eq. (298) of the Lecture Notes. To do this, just replace  $\mu_n$  in Eq. (255) with  $\mu_{eff}$  given by Eq. (298). Use a value of  $K$  such that  $K C_{ox} / (2\epsilon_s) = 0.5 \text{ V}^{-1}$ .

**Solution:**

The drain current is shown in the figure above by red dashed lines.



5. **Problem:**

Consider a Si  $n$ MOSFET with the following parameters:

substrate doping ( $p$ -type):  $N_A = 5 \times 10^{16} \text{ cm}^{-3}$

channel length  $L = 1.0 \text{ } \mu\text{m}$

gate width  $W = 10 \text{ } \mu\text{m}$

electron mobility  $\mu_n = 600 \text{ cm}^2/\text{Vs}$

oxide thickness  $t_{ox} = 4 \text{ nm}$

threshold voltage  $V_{T0} = 1 \text{ V}$ .

Using, as in the previous problem, the simplified model of the Lecture Notes, pages 159-162, especially Eq. (255), calculate the width of a similar  $p$ MOSFET giving the same saturated current for the same gate overdrive (that is, at the same value of  $|V_G - V_{T0}|$ ). Assume for the hole mobility a value of  $\mu_p = 200 \text{ cm}^2/\text{Vs}$ . If you wish, you may assume also for the threshold voltage of the  $p$ MOSFET the value of  $-1 \text{ V}$ . Recall that in dealing with the  $p$ FET all polarities ( $V_D$  and  $V_G$ ) are switched.

Do you really need to know the entire set of parameters characterizing the devices in order to reach your answer?

**Solution:**

No, you don't. Since the threshold voltage has the same magnitude (just of opposite polarity), the only reason why the two currents (for the  $n$  and  $p$  FETs) differ at a given gate overdrive is due to the different mobilities for the two devices. Since the hole mobility is three times smaller than the electron mobility, the  $p$ -channel FET must be three times wider, that is  $W = 30 \text{ } \mu\text{m}$ , in order to carry the same current.

